

A. Voltage References

1) Perform a “quick” analysis (e.g. assume 0.7V for a diode) to compute V_{REF} and current for each circuit in Fig. 1.

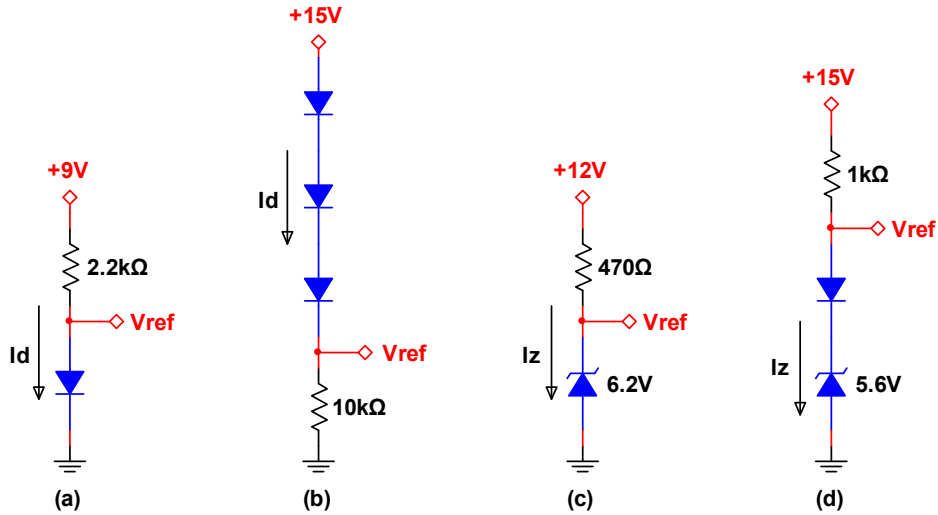


Fig. 1: Circuits to analyze for Problem 1.

2) In reality, the forward voltage of a diode is not always $V_F = 0.7V$. How to determine a more realistic value for V_F ? We need to use the diode’s data sheet! Fig. 2a shows the typical I-V curve for a 1N4148 diode. Notice that current is shown over a log scale, so the plot has a different shape compared to the sketch in the Lecture 01 notes (page 1.4).

- Perform a “quick” analysis of the circuit in Fig. 2b to determine I_d .
- Use the I-V curve in Fig. 2a to find the diode voltage V_F corresponding to your I_d value from part (a).
- Use your “new” V_F value from part (b) to analyze the diode circuit to determine the “new” I_d . In theory, you could repeat this procedure for finer accuracy, but it’s usually not necessary when doing hand calculations.

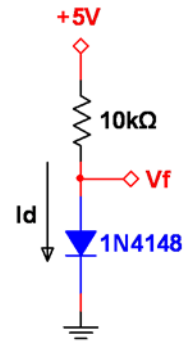
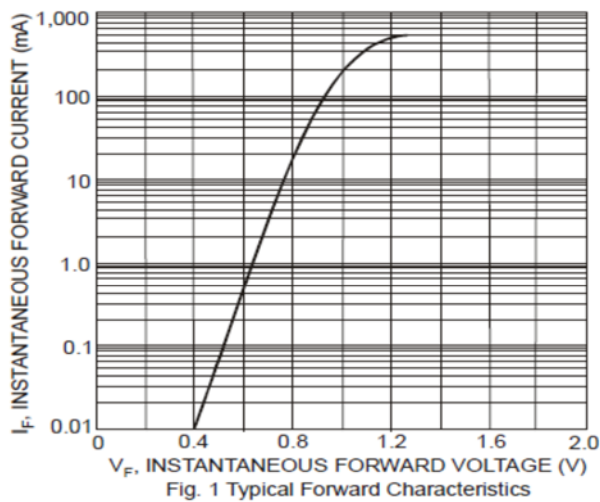


Fig. 2: Diode circuit and I-V curve from the 1N4148 datasheet.

- 3) Choose the appropriate resistor R (value and power rating) for a voltage reference that uses a 5.6V Zener in series with a diode (see Fig. 3). The power supply V_S can vary from 10 to 15V. For this problem, you can assume the diode and zener always have $V_F = 0.7V$ and $V_Z = 5.6V$, respectively. However, make sure the Zener has no less than 10 mA of current.

NOTE: It turns out this combination of components is very stable with temperature (e.g. $< 0.1\%$ per $^{\circ}C$). This is because the 5.6V Zener has the OPPOSITE temperature coefficient compared to the diode. You can buy a component that combines these two (e.g. 1N821).

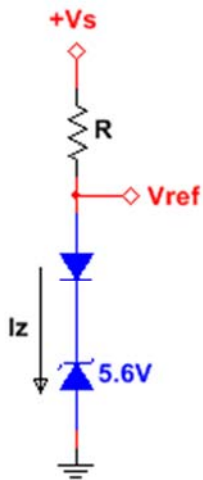


Fig. 3

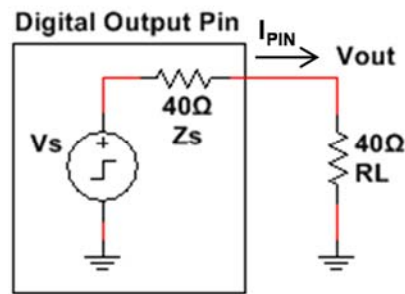


Fig. 4a

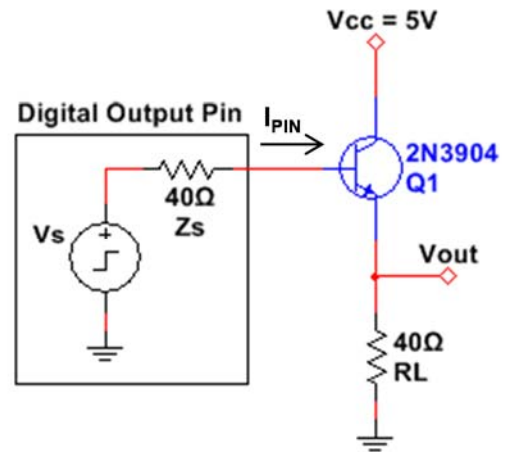


Fig. 4b

B. Emitter Follower Analysis (all data sheets on course website)

- 4) The digital output of a microcontroller (e.g. Arduino) typically produces 0 or 5V, but only if the output current is less than a certain value (e.g. less than 40 mA). The output voltage will drop at higher currents. ☹

Consider Fig. 4a, where the digital output pin is modeled as an ideal source $V_S = 5V$ in series with an output impedance $Z_S = 40\ \Omega$. Suppose we want to connect a $R_L = 40\ \Omega$ load to the digital output pin of a microcontroller. It is desirable for $V_{OUT} \geq 4V$ and $I_{PIN} \leq 40\ \text{mA}$.

- Compute the voltage V_{OUT} and current I_{PIN} coming out of the digital output pin.
- Is V_{OUT} acceptable?

Emitter follower to the rescue! As shown in Fig. 4b, a 2N3904 transistor is used for Q1. **Appendix #1 in the Lecture 01 notes discusses how to use the datasheet plots to find “typical” and “worst-case” transistor parameters.**

- Perform a “quick” analysis (e.g. assume $V_{BE} = 0.7V$ and $\beta = 100$) to determine if V_{OUT} and I_{PIN} are acceptable. Hint: Use KVL around the loop of V_S , Z_S , V_{BE} , and R_L .
- Compute V_{OUT} and I_{PIN} using **typical** 2N3904 parameters. Assume room temperature ($T = 25^{\circ}C$) for all parameters.
- Are V_{OUT} and I_{PIN} acceptable under “worst-case” 2N3904 parameters?

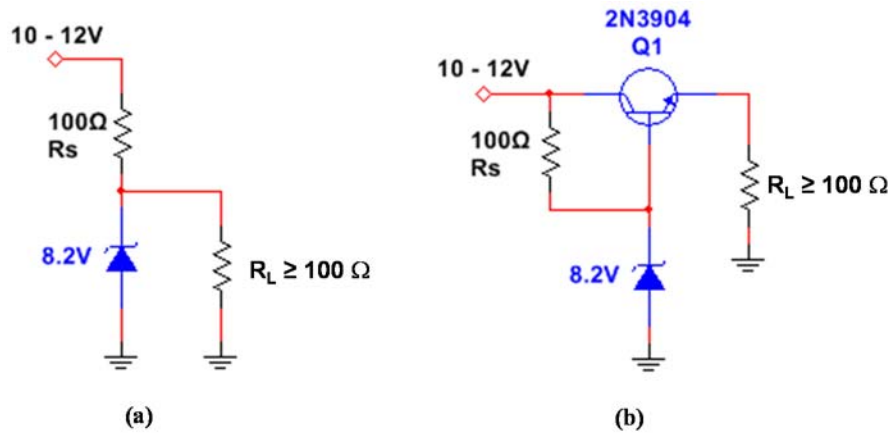


Fig. 5: (a) Zener directly connected to load. (b) Avoiding drop-out by using a Zener follower as a voltage buffer.

- 5) The output voltage of any DC power supply is based on an internal “voltage reference”, which is typically a zener diode. The main thing with a zener diode is to make sure it has enough current I_Z to avoid drop-out! In ECE 248, we assumed zener regulation is maintained if $I_Z > 0$ mA. That is not very realistic. **In this course, we’ll use $I_Z \geq 10$ mA as a more reasonable criterion for zener regulation.** There is no universally accepted rule for the minimum value of I_Z . Such ambiguity is often the case in the real world!

Suppose we want an 8.2V zener to drive loads with $R_L \geq 100$ ohm. The supply voltage varies from 10 to 12V.

- In Fig. 5a, does the zener diode maintain regulation under worst-case supply and load conditions?
 - In Fig. 5b, perform a “quick” analysis (assume $V_{BE} = 0.7V$ and $\beta = 100$) to determine if the zener diode maintains regulation under worst-case supply and load conditions.
 - Suppose Q1 is a 2N3904 transistor. Does Q1 need a heat sink? Assume “typical” transistor parameters (you need to use the datasheet plots).
 - We obviously want to avoid drop-out (I_Z too low) but we also want to avoid *burn-out* (I_Z too high)! Suppose the zener diode has a power rating of 1W. Is our zener OK?
- 6) If you need to control a large current with a very tiny current, use a Darlington! A brief review of Darlington transistors is in Appendix #3 of the Lecture 01 notes (see course website). The main thing to remember is that $V_{BE} \approx 1.4V$ and $\beta \approx 2500$ for a Darlington!

Fig. 6 shows a zener regulator consisting of a 12V zener, a TIP120 Darlington, and a load $R_L \geq 10$ ohm.

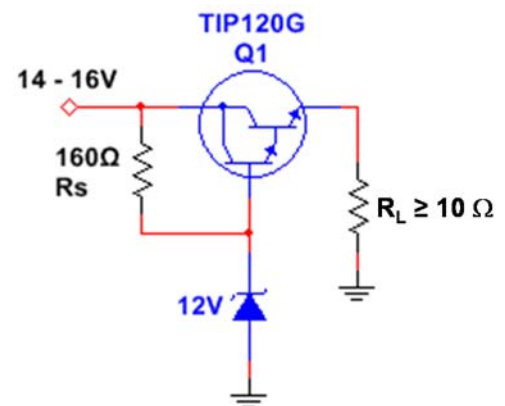


Fig. 6: The extra high current gain of a Darlington is useful for high load currents.

- Perform a “quick” Darlington analysis ($V_{BE} = 1.4V$, $\beta = 2500$) to determine if the zener diode maintains regulation under worst-case supply and load conditions.
- Assuming “quick analysis” conditions, compute the junction temperature T_J to show why Q1 very much needs a heat sink. Hint: You should find that T_J is over $350^\circ C$! Whoa.
- You have a selection of heat sinks with sink-to-ambient thermal resistance $\theta_{SA} = 5, 10, 15,$ and 20 $^\circ C/W$. Which heat sink is the best choice? Assume the case-to-sink thermal resistance is $\theta_{CS} = 0.5^\circ C/W$ (see Appendix #4 in the Lecture 01 notes for heat sink formulas). Assume $T_A = 25$ $^\circ C$ and that we want $T_J < 85^\circ C$.

7) When choosing a transistor without a heat sink, we want the device's power rating to be over twice the actual power dissipation (i.e. 2X safety factor). However, the power rating depends on the ambient temperature T_A ! Most data sheets show the power rating for $T_A = 25\text{ }^\circ\text{C}$, which is OK if the device is in the open air. In many situations, the circuit is inside an enclosure without a fan (e.g. smartphone), so the ambient temperature of the transistor can easily exceed $T_A = 40\text{ }^\circ\text{C}$! This reduction (also called "derating") of the device's power rating is important.

(a) What is the 2N3904's power rating at $T_A = 25, 50,$ and $75\text{ }^\circ\text{C}$ without a heat sink?

(b) A TIP31 can easily handle 750 mW at $T_A = 25\text{ }^\circ\text{C}$. But can it do so at $T_A = 75\text{ }^\circ\text{C}$ without a heat sink?

For this course, we'll choose a heat sink by requiring $T_J < 85\text{ }^\circ\text{C}$, which is comparable to a cup of scalding hot coffee! This produces similar results to the "2X safety factor" approach (e.g. what we use for a transistor WITHOUT a heat sink). If your calculation produces a NEGATIVE value for θ_{SA} , this is obviously impossible and a heftier transistor is required!

(c) A transistor must dissipate 15W. Can a TIP31 do this with a heat sink for $T_A = 40\text{ }^\circ\text{C}$? Assume $\theta_{CS} = 0.5^\circ\text{C/W}$. Show all work!

(d) Can a MJE3055 dissipate 15W at $T_A = 40\text{ }^\circ\text{C}$? If so, compute the minimum heat sink (i.e. maximum θ_{SA}) required.

C. Emitter Follower Design (all data sheets on course website)

8) Design a zener follower (see Fig. 8) for a load rated at 6V @ 50 mA. The input voltage V_S is 7 to 9V. Some design constraints:

- Output voltage: The load is OK between 5.5 to 6.5V
- Zener: Choose from 4.7, 5.1, 5.6, 6.2, 6.8, or 7.5V
- Transistor: Choose from 2N3904, TIP31, MJE3055

(a) Using "quick" analysis assumptions ($V_{BE} = 0.7\text{V}$ and $\beta = 100$), determine the appropriate zener diode. Show all work!

(b) Determine the appropriate transistor using "quick" analysis parameters (assume $T_A = 25\text{ }^\circ\text{C}$). You MUST explain why you chose one transistor and not the other two.

NOTE: Remember that you want to choose a transistor of appropriate size. As a circuit designer, you do not want your components to be larger or more expensive than necessary!

(c) Choose the appropriate 5% standard value for R_S , and determine if a $\frac{1}{4}\text{ W}$, $\frac{1}{2}\text{ W}$, or 1 W rating is needed.

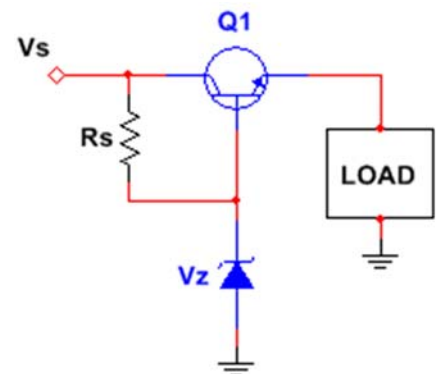


Fig. 8: Circuit for Problem 8.

9) You are asked to design a power supply for a 6V DC motor. The input voltage V_S comes from a supply that can vary from 9V to 11V. Normally, you would use a DC-DC converter chip for this. Unfortunately, there seems to be a world-wide shortage of such ICs. You decide to use a zener regulator with a Darlington, due to the potentially large motor current (see below).

- Here are the design constraints:

- Output voltage: must be within +/- 5% of 6V.
- Motor: The motor has a “no-load” current of 150 mA and a “stall” current of 1A. Don’t design for the no-load current, because that assumes the motor is not connected to anything – NOT useful! Let’s design the power supply to provide 50% of the stall current.
- Zener: The zener must either be 6.2, 6.8, 7.5, or 8.2V.
- The Darlington transistor can either be a KSP13, TIP120, or TIP140.
- $T_A = 40\text{ }^\circ\text{C}$ (circuit is in an enclosure, where the air gets warm)

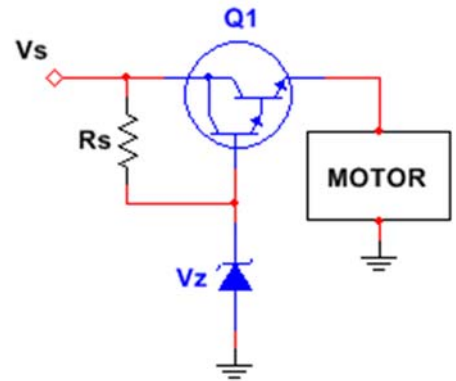


Fig. 9: Darlington zener follower.

- (a) Using “quick” analysis assumptions, determine the appropriate zener diode. Show all work!
- (b) Perform a “quick” analysis to determine the appropriate transistor. If you need a heat sink, leave those calculations for part (c). You MUST explain why you chose one transistor and not the other two.
- (c) The available heat sinks have thermal resistances of $\theta_{SA} = 6, 12, 18,$ and $24\text{ }^\circ\text{C/W}$. Which heat sink is the best choice? Assume that $T_J < 85^\circ\text{C}$ and $\theta_{CS} = 0.5^\circ\text{C/W}$. Remember that $T_A = 40\text{ }^\circ\text{C}$! Show all work!
- (d) Using “**typical**” parameters for your choice of transistor from part (b), determine the appropriate resistor R_S . Choose a standard 5% value and a power rating of either $\frac{1}{4}\text{ W}$, $\frac{1}{2}\text{ W}$, or 1 W .

D. Emitter Follower with Feedback

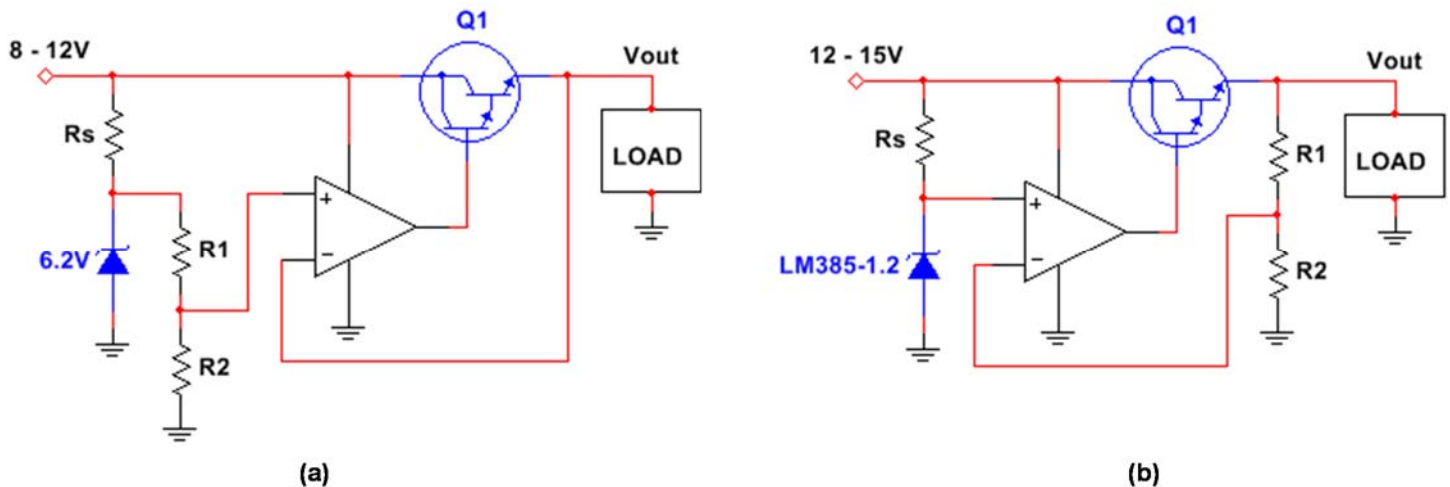


Fig. 10: Linear voltage regulator with (a) $V_{OUT} < V_{REF}$ (b) $V_{OUT} > V_{REF}$.

- 10) A voltage regulator uses negative feedback to stabilize the output voltage. A voltage divider is used to sample a portion of either V_{REF} or V_{OUT} , depending on whether V_{OUT} is smaller or larger than V_{REF} .
 - (a) In Fig. 10a, V_{REF} is a 6.2V zener while $V_{OUT} = 5\text{ V}$ (+/- 5% is OK). **Choose the standard 5% resistor values and power ratings ($\frac{1}{4}\text{ W}$, $\frac{1}{2}\text{ W}$, or 1 W) for R_1 , R_2 , and R_S .** Keep in mind the following:
 - The voltage divider should draw less than 1 mA ($< 10\%$ of the zener current) to ensure the zener is little affected by the divider.
 - The zener should have at least 10 mA.

(b) In Fig. 10b, V_{REF} is an LM385-1.2 “bandgap voltage reference” that produces 1.235V while $V_{OUT} = 9V$ (+/- 5% is OK). **Choose the standard 5% resistor values and power ratings ($\frac{1}{4} W$, $\frac{1}{2} W$, or $1W$) for R_1 , R_2 , and R_s .** Keep in mind the following:

- Let the LM385-1.2 have about 1 mA (this is much lower than a conventional zener!).

NOTE: The LM385-1.2 is really nice because it works even with only 10 μA of current! This makes it ideal for micropower (e.g. long battery life) applications.

- The voltage divider should draw about 1 mA. This is enough current to keep the pass transistor turned on but low enough to ensure the divider dissipates little power.