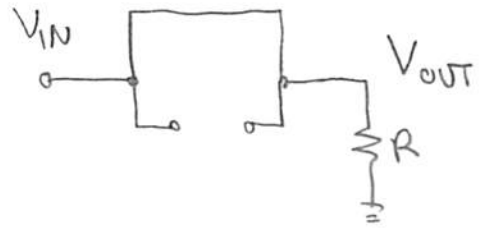
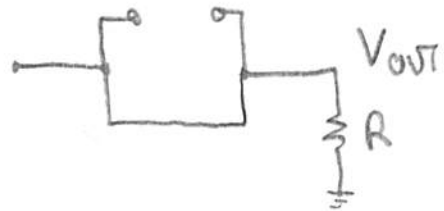


① a) $f=0$: Inductor \rightarrow short
Capacitor \rightarrow open



$f \rightarrow \infty$: Inductor \rightarrow open
Capacitor \rightarrow short



b) Due to short, $V_{OUT} = V_{IN} \rightarrow \boxed{\frac{V_{OUT}}{V_{IN}} = 1}$

c) $Z = j\omega L \parallel \frac{1}{j\omega C}$

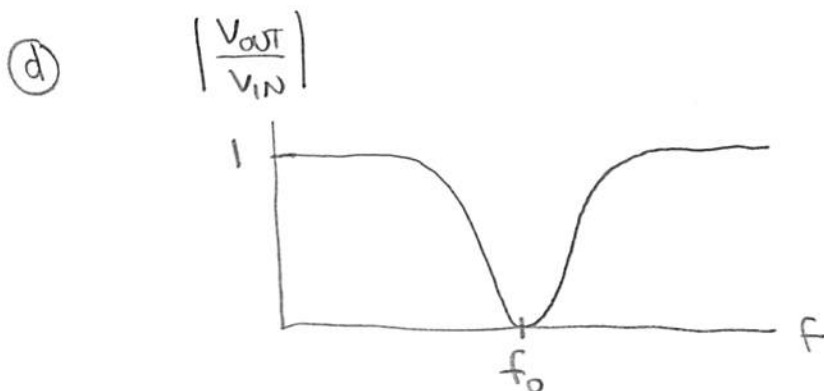
$$= \frac{j\omega L \frac{1}{j\omega C}}{j\omega L + \frac{1}{j\omega C}} = \frac{j\omega L}{-\omega^2 LC + 1} = \frac{j\omega L}{1 - \omega^2 LC}$$

Looks like OPEN ∇ at $f=f_0$.

$\rightarrow 1 - \omega_0^2 LC = 0$

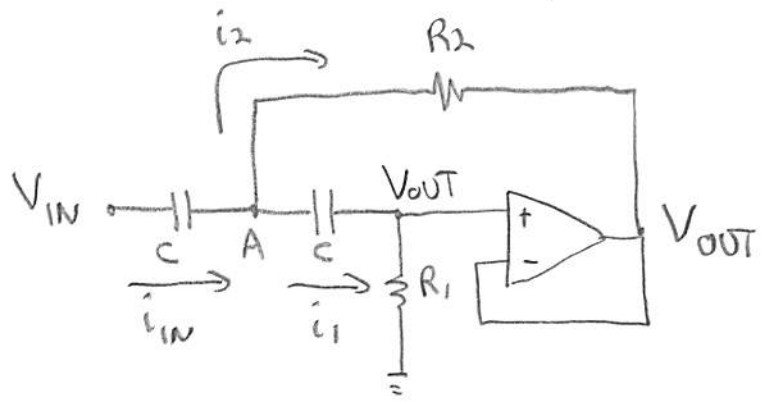
$\omega_0 = \frac{1}{\sqrt{LC}} \rightarrow \boxed{f_0 = \frac{1}{2\pi\sqrt{LC}}}$

$|Z| \rightarrow \infty$ when denominator = 0



Notch filter
 \uparrow
or band stop

②



ⓐ KCL at node A:

$$i_{IN} = i_1 + i_2$$

$$\frac{V_{IN} - V_A}{Z} = \frac{V_A - V_{OUT}}{Z} + \frac{V_A - V_{OUT}}{R_2}$$

$\swarrow \frac{1}{j\omega C} \searrow$

$$V_A \cdot i_1 = \frac{V_A - V_{OUT}}{Z} = \frac{V_{OUT} - 0}{R_1}$$

$$\frac{R_1}{Z} V_A = \left(1 + \frac{R_1}{Z}\right) V_{OUT}$$

$$V_A = \left(1 + \frac{Z}{R_1}\right) V_{OUT} \quad (2)$$

$$V_A - V_{OUT} = \frac{Z}{R_1} V_{OUT} \quad (3)$$

$$V_{IN} - V_A = V_A - V_{OUT} + \frac{Z}{R_2} (V_A - V_{OUT})$$

$$V_{IN} - V_A = \left(1 + \frac{Z}{R_2}\right) (V_A - V_{OUT}) \quad (1)$$

Plug (2) and (3) into (1):

$$V_{IN} - \left(1 + \frac{Z}{R_1}\right) V_{OUT} = \left(1 + \frac{Z}{R_2}\right) \frac{Z}{R_1} V_{OUT}$$

$$V_{IN} = \left[\left(1 + \frac{Z}{R_2}\right) \frac{Z}{R_1} + \left(1 + \frac{Z}{R_1}\right) \right] V_{OUT}$$

$$\frac{V_{OUT}}{V_{IN}} = \frac{1}{\frac{Z}{R_1} + \frac{Z^2}{R_1 R_2} + 1 + \frac{Z}{R_1}} = \frac{1}{1 + 2 \frac{Z}{R_1} + \frac{Z^2}{R_1 R_2}}$$

$$= \frac{1}{1 + 2 \frac{1}{j\omega R_1 C} + \frac{1}{-\omega^2 R_1 R_2 C^2}} = \frac{-\omega^2 R_1 R_2 C^2}{-\omega^2 R_1 R_2 C^2 \left(1 - j \frac{2}{\omega R_1 C}\right) + 1}$$

$$= \frac{-\omega^2 R_1 R_2 C^2}{1 + j 2 \omega R_1 C - \omega^2 R_1 R_2 C^2}$$

$$\textcircled{b} \quad -\omega^2 \underbrace{R_1 R_2 C^2}_{\frac{1}{\omega_p^2}} = -\omega^2 / \omega_p^2 \rightarrow \boxed{\omega_p = \frac{1}{\sqrt{R_1 R_2 C}}}$$

Also, $2\cancel{\omega} R_2 C = \frac{\omega}{\omega_p Q} \rightarrow Q = \frac{1}{\omega_p 2 R_2 C} = \frac{\sqrt{R_1 R_2} \cancel{C}}{2 R_2 \cancel{C}}$

$$\boxed{Q = \frac{1}{2} \sqrt{\frac{R_1}{R_2}}}$$

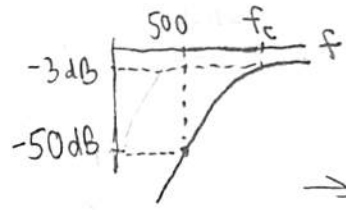
\textcircled{c} When $Q = \frac{1}{\sqrt{2}}$ and $f_c = \frac{1}{2\pi} \omega_p$

$$\rightarrow \frac{V_{out}}{V_{in}} = \frac{-f^2/f_p^2}{1 + j \frac{f}{f_p} \sqrt{2} - f^2/f_p^2}$$

$$\left| \frac{V_{out}}{V_{in}} \right| = \frac{f^2/f_p^2}{\sqrt{(1 - f^2/f_p^2)^2 + (\frac{f}{f_p} \sqrt{2})^2}} = \frac{(f/f_p)^2}{\sqrt{1 - 2(f/f_p)^2 + (f/f_p)^4 + 2(f/f_p)^2}}$$

$$\boxed{\left| \frac{V_{out}}{V_{in}} \right| = \frac{(f/f_p)^2}{\sqrt{1 + (f/f_p)^4}}}$$

3) (a) $20 \log_{10} |A| = -50 \text{ dB} \Rightarrow |A| = 10^{-5/20} = 0.00316 = \frac{(500/f_c)^2}{\sqrt{1 + (500/f_c)^4}}$



$$\Rightarrow 10^{-5} = \frac{(500/f_c)^4}{1 + (500/f_c)^4} \Rightarrow 1 + \left(\frac{500}{f_c}\right)^4 = 10^5 \left(\frac{500}{f_c}\right)^4$$

$$1 = (10^5 - 1) \left(\frac{500}{f_c}\right)^4$$

$$1 = (10^5 - 1)^{1/4} \frac{500}{f_c}$$

$$\Rightarrow \boxed{f_c = 8891.4 \text{ Hz}}$$

$$f_c = \frac{1}{2\pi \sqrt{R_1 R_2} C}, \quad Q = \frac{1}{2} \sqrt{\frac{R_1}{R_2}} = \frac{1}{\sqrt{2}} \Rightarrow \frac{R_1}{R_2} = 2$$

$$\Rightarrow f_c = \frac{1}{2\pi \sqrt{2 R_2^2} C} = \frac{1}{2\pi \sqrt{2} R_2 C} = 8891.4 \text{ Hz}$$

$$R_2 C = \frac{1}{2\pi \sqrt{2} (8891.4)} = 1.266 \times 10^{-5} \text{ s}$$

$$\text{Let } \boxed{C = 1 \text{ nF}} \Rightarrow R_2 = 12657 \Omega$$

Want $R_1, R_2 \geq k\Omega$
To reduce power
dissipation

Choose $\boxed{R_2 = 13 \text{ k}}$

$\boxed{R_1 = 26 \text{ k}}$



$$\Rightarrow \text{Actual } f_c = \frac{1}{2\pi \sqrt{26 \times 10^3 \times 13 \times 10^3 \times 10^{-9}}}$$

$$= 8656.9 \text{ Hz}$$

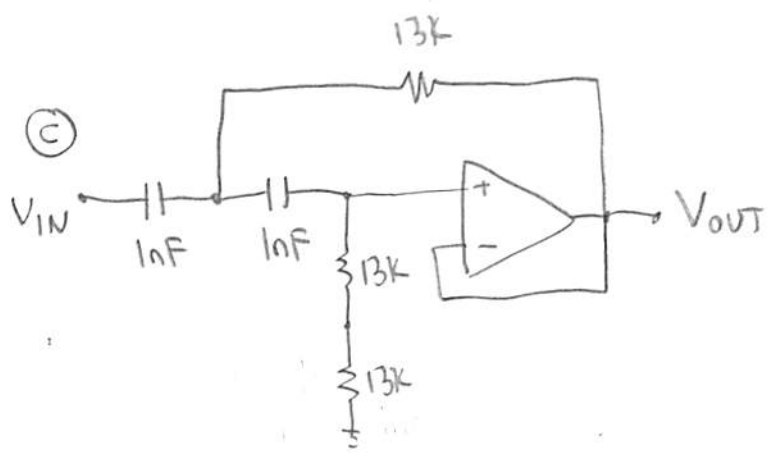
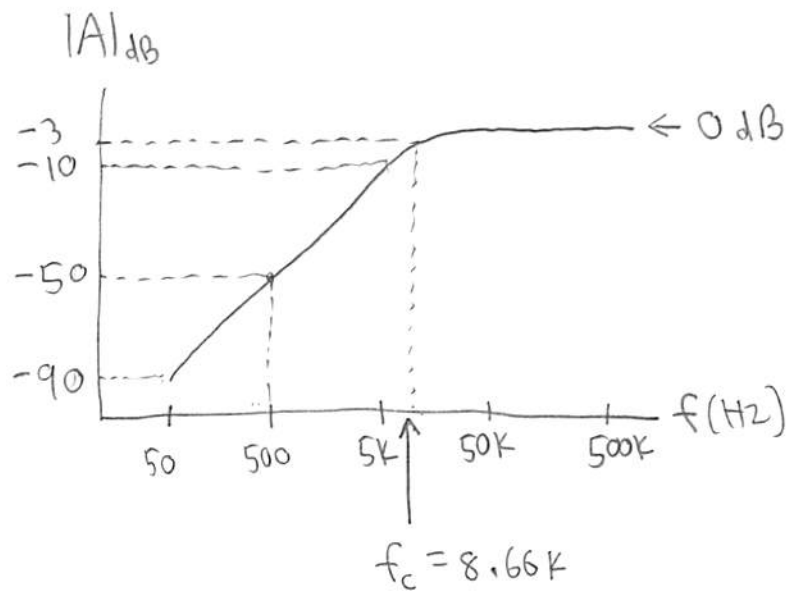
within 1 dB!

$$|A| = \frac{(500/8656.9)^2}{\sqrt{1 + (500/8656.9)^4}} = 0.003336 \Rightarrow \boxed{-49.54 \text{ dB}}$$

(b) $f_c = 8656.9 \text{ Hz}$

$|A| \sim -50 \text{ dB}$ at 500 Hz

2nd order filter rolloff is -40 dB/decade



(4)

(a) $Q_1 = 0.54 = \frac{1}{2} \sqrt{\frac{C_2}{C_1}} \rightarrow (1.08)^2 = \frac{C_2}{C_1} \rightarrow C_2 = 1.1664 C_1$

$f_c = \frac{1}{2\pi R \sqrt{C_1 C_2}} = 20 \times 10^3$

Try $C_1 = 1000 \text{ pF}$
 $C_2 = 1200 \text{ pF}$

$R = \frac{1}{2\pi \sqrt{1000 \times 10^{-12} \times 1200 \times 10^{-12}} \times 20 \times 10^3} = 7264.4 \Omega$

$\rightarrow \text{Actual } f_c = \frac{1}{2\pi (7.5 \times 10^3) \sqrt{1000 \times 1200 \times 10^{-24}}} = 19371.7 \text{ Hz}$
 $100 \times \frac{19371.7 - 20000}{20000} = -3.1\%$

(b) $Q_2 = 1.31$

$(2.62)^2 = \frac{C_2}{C_1}$

$C_2 = 6.8644 C_1$

Try $C_2 = 6800 \text{ pF}, C_1 = 1000 \text{ pF}$

$R = \frac{1}{2\pi \sqrt{6800 \times 1000 \times 10^{-12} \times 20 \times 10^3}} = 3051.6 \Omega$

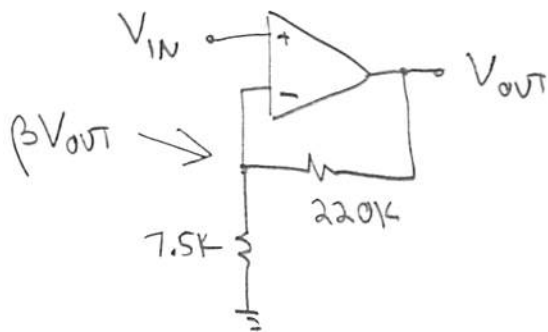
Actual f_c : $\frac{1}{2\pi \sqrt{6800 \times 1000 \times 10^{-12} \times 2000}} = 20,344 \text{ Hz}$
 $\Delta = 1.7\%$

5

a)

$$\beta V_{out} = \frac{7.5K}{7.5K + 220K} V_{out}$$

$$\Rightarrow \boxed{\beta = 0.03297}$$



b)

$$A_o = 106 \text{ dB} \rightarrow A_o = 10^{106 \text{ dB} / 20 \text{ dB}} = 1.995 \times 10^5$$

$$\Rightarrow \boxed{A_o \beta = 6577}$$

c)

$$G = \frac{A}{1 + A\beta} = \frac{1.995 \times 10^5}{1 + 6577} = \boxed{30.328}$$

d)

$$Z_{in} = (1 + A\beta) R_i = (1 + 6577)(200K) = \boxed{1.316 \text{ G}\Omega}$$

e)

$$Z_{out} = \frac{R_o}{1 + A\beta} = \frac{25}{1 + 6577} = 3.8 \times 10^{-3} \Omega = \boxed{3.8 \text{ m}\Omega}$$

6

$$\text{a) want } G = 100 = \frac{A_o}{1 + A_o \beta} \rightarrow 1 + A_o \beta = \frac{A_o}{100}$$

$$\beta = \frac{\frac{A_o}{100} - 1}{A_o} = \frac{1}{100} - \frac{1}{A_o}$$

$$\text{Since } A_o = 66 \text{ dB} = 10^{66 \text{ dB} / 20 \text{ dB}} = 1995.3$$

$$\rightarrow \beta = 0.01 - \frac{1}{1995.3} = \boxed{0.00950}$$

NOTE: Only A_o varies while β is constant!

(b) $\frac{\Delta G}{G} = \frac{G_{MAX} - G_{MIN}}{G_o} \times 100\%$

Max: $A_{o,MAX} = 10^{69/20} = 2818.4 \Rightarrow G_{MAX} = \frac{2818.4}{1 + (2818.4)(.0095)} = \underline{\underline{101.47}}$

Min: $A_{o,MIN} = 10^{63/20} = 1412.5 \Rightarrow G_{MIN} = \frac{1412.5}{1 + (1412.5)(.0095)} = \underline{\underline{97.96}}$

$\Rightarrow \frac{\Delta G}{G} = \frac{101.47 - 97.96}{100} \times 100\% = \boxed{3.5\%}$

Does NOT satisfy design spec.

(c) Want $G_1 \times G_2 = G^2 = 100 \Rightarrow G = 10 = \frac{A_o}{1 + A_o \beta}$

$1 + A_o \beta = 0.1 A_o$

$\beta = 0.1 - \frac{1}{A_o} = 0.1 - \frac{1}{1995.3}$

$\boxed{\beta = 0.09950}$

(d) $G_{MAX}^2 = \left(\frac{A_{MAX}}{1 + A_{MAX} \beta} \right)^2 = \left(\frac{2818.4}{1 + 2818.4 \times .0995} \right)^2 = 100.29$

$G_{MIN}^2 = \left(\frac{A_{MIN}}{1 + A_{MIN} \beta} \right)^2 = \left(\frac{1412.5}{1 + 1412.5 \times .0995} \right)^2 = 99.59$

$\Rightarrow \frac{\Delta G}{G} = \frac{100.29 - 99.59}{100} \times 100\% = \boxed{0.7\%}$



7

⑦ a) $G = \frac{A}{1+A\beta} \left\{ \begin{array}{l} G_{MAX} = \frac{A_{MAX}}{1+A_{MAX}\beta} = \frac{10A_{MIN}}{1+10A_{MIN}\beta} \\ G_{MIN} = \frac{A_{MIN}}{1+A_{MIN}\beta} \end{array} \right.$

We know $0.99 G_0 < G < 1.01 G_0$

$$G_{MAX} = 1.01 \times 30 = 30.3 = \frac{10A_{MIN}}{1+10A_{MIN}\beta} \quad (1)$$

$$G_{MIN} = 0.99 \times 30 = 29.7 = \frac{A_{MIN}}{1+A_{MIN}\beta} \quad (2)$$

We have two equations and two unknowns (A_{MIN} and β)

Using (2): $A_{MIN}\beta = \frac{A_{MIN}}{29.7} - 1$

Plug into (1): $30.3 = \frac{10A_{MIN}}{1+10\left[\frac{A_{MIN}}{29.7} - 1\right]} = \frac{10A_{MIN}}{10\frac{A_{MIN}}{29.7} - 9}$

$$\frac{10}{29.7} A_{MIN} - 9 = \frac{10}{30.3} A_{MIN}$$

$$A_{MIN} = \frac{9}{\frac{10}{29.7} - \frac{10}{30.3}} = 1349.9 \sim \boxed{1350}$$

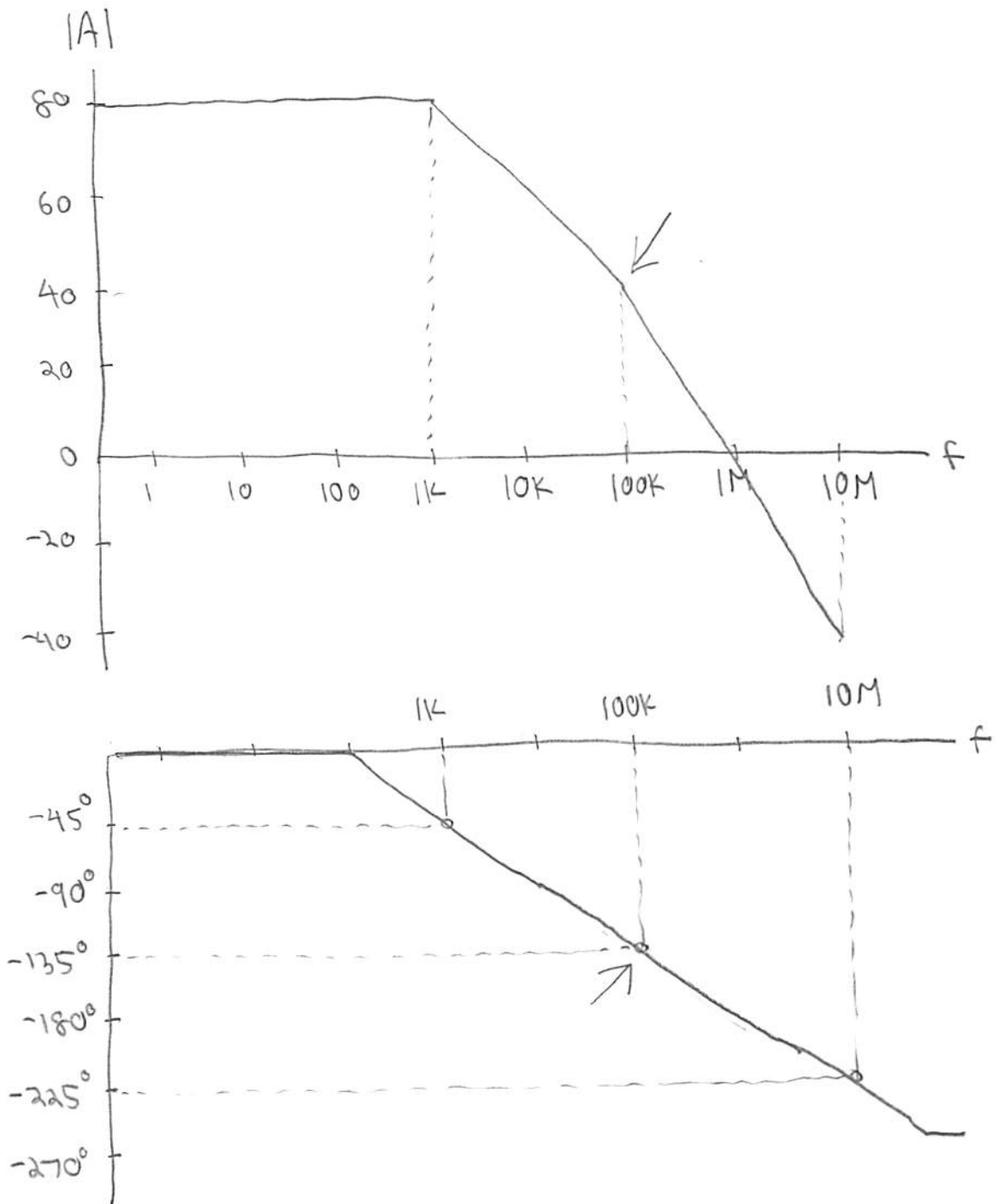
⑧ Using (2): $A_{MIN}\beta = \frac{A_{MIN}}{29.7} - 1 \Rightarrow \beta = \frac{\frac{1350}{29.7} - 1}{1350} = \boxed{.03293}$

⑨ $G_0 = 30 = \frac{A_0}{1+A_0\beta} \Rightarrow 1+A_0\beta = \frac{A_0}{30} \Rightarrow 1 = A_0\left(\frac{1}{30} - \beta\right)$

$$A_0 = \frac{1}{\frac{1}{30} - .03293} = \boxed{2479}$$

⑧ a

$$20 \log_{10} 10^4 = 80 \text{ dB}$$



⑥ $45^\circ \text{ phase margin} = \angle A - (-180^\circ) \Rightarrow \angle A = -135^\circ \leftarrow f = f_{p2} = 100 \text{ kHz}$

From Bode plot of $|A|$: $|A| = 40 \text{ dB} @ 100 \text{ kHz}$

$$= 1/\beta \rightarrow 1/\beta = 10^{40/20} = 100$$

$$\Rightarrow \boxed{\beta = 0.01}$$

⊖ $\beta=1 \rightarrow 20 \log_{10}(1) = 0 \text{ dB}$



$$\angle A = -\tan^{-1}\left(\frac{10^6}{10^3}\right) - \tan^{-1}\left(\frac{10^6}{10^5}\right) - \tan^{-1}\left(\frac{10^6}{10^7}\right)$$

$$= -89.94^\circ - 84.29^\circ - 5.71^\circ = \boxed{-179.94^\circ} \leftarrow$$

Very close to -180° !
 ☹️

ⓓ NO, phase margin is $.06^\circ$ when $\beta=1$!